

# Radar Imaging Homework 1

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## Introduction

We are given a system with the scenario depicted in Figure 1. A UAV radar illuminates a scene with three targets placed in different ground range positions.

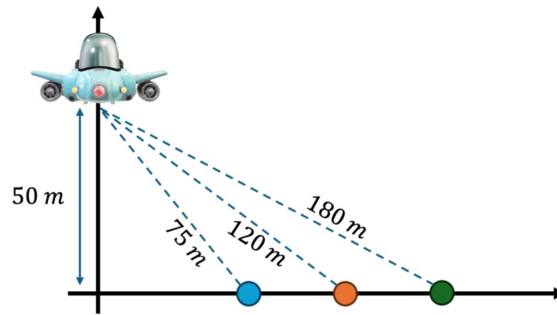


Figure 1: System Geometry

## Given Information

It is given that all the targets are placed at boresight for simplicity. So,  $f_{tx} = f_{rx} = 1$ . All the other system parameters are mentioned in the table present in the material we were provided.

*Note: Transmitted power ( $P_{TX}$ ) mentioned in the table is used only in the solution of question 4.*

## 1 Derivation of the transmitted power

### 1.1 From Given Information

We are given the information that the signal-to-noise-ratio before range compression is given to be 0 dB. We also have to consider the target at the range of 180 m with a  $\sigma = 15 \text{ m}^2$ .

$$T_{\text{sys}} = T_{\text{scene}} + T_{\text{rec}} = 578.6261 \text{ [K]}$$

$$\Delta\theta_{\text{rad}} = 40^\circ \cdot \frac{\pi}{180} \approx 0.698 \text{ [rad]}$$

$$\Delta\psi_{\text{rad}} = 40^\circ \cdot \frac{\pi}{180} \approx 0.698 \text{ [rad]}$$

### 1.2 To Derive

We are asked to derive the transmitted power in dBm for the above mentioned requirements.

### 1.3 Derivation

#### Antenna Gain and Effective Area

Approximate solid angle:

$$\Omega_A = \Delta\theta_{\text{rad}} \times \Delta\psi_{\text{rad}} \approx 0.698 \times 0.698 \approx 0.487 \text{ [sr]} \quad (1)$$

Gain:

$$G \approx \frac{4\pi}{\Omega_A} \approx \frac{4\pi}{0.487} \approx 25.8 \quad (2)$$

Effective Aperture:

$$A_e = \frac{\lambda^2 G}{4\pi} = \frac{(0.03)^2 \cdot 25.8}{4\pi} \approx 1.847 \times 10^{-3} \text{ [m}^2\text{]} \quad (3)$$

#### Incident Power Density

$$S_i = \frac{P_{\text{Tx}} G_{\text{Tx}} f_{\text{Tx}}(\theta_0, \psi_0)}{4\pi R_0^2} \quad (4)$$

Solving for transmit power:

$$P_{\text{Tx}} = \frac{S_i 4\pi R_0^2}{G_{\text{Tx}} f_{\text{Tx}}(\theta_0, \psi_0)} \quad (5)$$

#### Scattered Power

$$P_s = S_i \sigma \quad (6)$$

Scattered power density at receiver:

$$S_r = \frac{P_s}{4\pi R_0^2} = \frac{S_i \sigma}{4\pi R_0^2} \quad (7)$$

#### Received Power

$$P_{\text{Rx}} = S_r A_e f_{\text{Rx}}(\theta_0, \psi_0) \quad (8)$$

Substitute Eq 7 in Eq 8:

$$P_{\text{Rx}} = \frac{S_i \sigma}{4\pi R_0^2} A_e f_{\text{Rx}}(\theta_0, \psi_0) \quad (9)$$

Solving Eq 9 for  $S_i$ :

$$S_i = \frac{P_{\text{Rx}} 4\pi R_0^2}{\sigma A_e f_{\text{Rx}}(\theta_0, \psi_0)}$$

#### Noise and SNR

Given SNR = 0 dB:

$$\text{SNR} = \frac{P_{\text{Rx}}}{P_N} = 1 \Rightarrow \text{SNR} = P_N = P_{\text{RX}} \quad (10)$$

Noise power:

$$P_N = P_{\text{RX}} = kT_{\text{sys}}B$$

Solving for  $P_N$ , we get,

$$P_N = 1.38 \times 10^{-23} \times 578.6261 \times 100 \times 10^6 \approx 7.9850 \times 10^{-13} \text{ [W]}$$

$$P_{\text{RX}} \approx 7.9850 \times 10^{-13} \text{ [W]} \quad (11)$$

Directivity function is given to be

$$f_{\text{RX}}(\theta_o, \phi_o) = f_{\text{TX}}(\theta_o, \phi_o)$$

$$f_{\text{TX}}(\theta_o) \cdot f_{\text{TX}}(\phi_o) = f_{\text{RX}}(\theta_o) \cdot f_{\text{RX}}(\phi_o) \quad (12)$$

We know the elevation of the radar is  $45^\circ$  and we need to find the elevation of the target at  $R_0 = 180\text{[m]}$ ,

$$\cos^{-1}(H/R_0) = \cos^{-1}(50/180) \approx 1.28\text{[rad]} \approx 73.87^\circ$$

Adjusting elevation after with the elevation of the radar we get,

$$\theta = 28.8^\circ \approx 0.50\text{[rad]} \quad (13)$$

We know the target is in boresight so we can assume  $\phi = 0^\circ$ .

Substituting for result obtained in Eq.13 in Eq.12 we get,

$$\begin{aligned} f_{TX}(\theta_o, \phi_o) &= \text{sinc}(\theta)^2 \times \text{sinc}(\phi)^2 \\ f_{TX}(\theta_o, \phi_o) &= \text{sinc}(28.8^\circ)^2 \times 1^2 \approx 0.633^2 \times 1 = 0.4053 \end{aligned} \quad (14)$$

$$\begin{aligned} S_i &= \frac{7.9850 \times 10^{-13} \times 4\pi \times 180^2}{15 \times 1.847 \times 10^{-3} \times 0.4053} \\ &= 2.8953 \times 10^{-5} \text{ [W]} \end{aligned} \quad (15)$$

$$\begin{aligned} P_{TX} &= \frac{2.8953 \times 10^{-5} \times 4\pi \times 180^2}{25.8 \times 0.4053} \\ &= 1.1273 \text{ [W]} \end{aligned} \quad (16)$$

$$\begin{aligned} &= 10 \log_{10}(1.1273) = 0.5205 \text{ dBW} + 30 \\ &= 30.5205 \text{ [dBm]} \end{aligned} \quad (17)$$

## 2 Matlab Implementation:

### 2.1 Given Information:

We are provided with the  $P_{tx}$  and all the other values needed to find the Signal-to-Noise Ratio(SNR).

### 2.2 To Implement:

We need to implement the equations from the provided material to calculate the SNR before range compression.

### 2.3 Implementation:

The implementation can be found in the attached file (Question2.m).

### 2.4 Results:

A screenshot of the obtain results after the execution of the Matlab file has been provided in the Figure 2

## 3 Derivation of SNR After Range Compression

### Given Information

We are provided with all the radar parameters.

### To solve:

We are asked to derive the SNR after range compression.

```

--- SNR Before Range Compression (Assuming ftx=1) ---
P_TX = 10.00 dBm (1.0000e-02 W)
T_sys = 578.6261 K
P_N = 7.9850e-13 W
Gtx = 25.78
A_e = 1.8466e-03 m^2
-----
Target 1 (R0=75m, sigma=1m2):
Directivity function : 0.9898
P_rx: 9.3363e-14 W
SNR: 1.1692e-01 (-9.32 dB)
-----
Target 2 (R0=120m, sigma=5m2):
Directivity function : 0.6473
P_rx: 3.0464e-14 W
SNR: 3.8152e-02 (-14.18 dB)
-----
Target 3 (R0=180m, sigma=15m2):
Directivity function : 0.3989
P_rx: 6.8567e-15 W
SNR: 8.5869e-03 (-20.66 dB)
-----

Final SNR (linear) array for [75m, 120m, 180m] targets:
0.1169 0.0382 0.0086

Final SNR (dB) array for [75m, 120m, 180m] targets:
-9.3210 -14.1849 -20.6616

```

Figure 2: SNR obtained from the theoretical calculations. [Question 2]

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Final SNR (linear) array, after range compression, for target [75, 120, 180]:
13.1405 11.7154 11.1223

Final SNR (dB) array, after range compression, for target [75, 120, 180]:
41.1861 40.6876 40.4620

```

Figure 3: SNR obtained from the Matlab simulation [Question 4]

## Derivation

The received power is derived by substitution in Equation 5:

$$P_{RX} = P_{TX} \frac{\sigma \lambda^2 G^2 f_{TX}^2}{(4\pi)^3 R^4} \quad (18)$$

The signal energy at the matched filter input is  $E_s = P_{RX} T_p$ , where  $T_p$  is the pulse duration. The matched filter output peak is  $|y_{sig}(t_0)| = E_s$ . The signal power at the output peak (amplitude squared) is:

$$P_{signal,after} = |y_{signal}(t_0)|^2 = (P_{RX} T_p)^2 \quad (19)$$

The power spectral density of the thermal noise is obtained by,

$$S_n(f) = N_0 = kT_{sys}$$

For a matched Filter the power of the output noise is obtained by using Fourier Transform,

$$P_{noise,after} = kT_{sys} \int |h(t)|^2 dt = kT_{sys} P_{RX} T_p \quad (20)$$

The output SNR is the ratio of  $P_{signal,after}$  to  $P_{noise,after}$ :

$$SNR_{after} = \frac{P_{signal,after}}{P_{noise,after}} = \frac{(P_{RX} T_p)^2}{kT_{sys} P_{RX} T_p} = \frac{P_{RX} T_p}{kT_{sys}} \quad (21)$$

Substituting  $P_{RX}$  from the first step:

$$SNR_{after}(R, \sigma) = \frac{P_{TX} \sigma \lambda^2 G^2 f_{TX}^2 T_p}{(4\pi)^3 R^4 kT_{sys}} \quad (22)$$

The SNR before compression is  $SNR_{before} = \frac{P_{RX}}{kT_{sys} B}$ , where  $B$  is the bandwidth.

Taking the ratio of  $SNR_{after}$  to  $SNR_{before}$ :

$$\frac{SNR_{after}}{SNR_{before}} = \frac{P_{RX} T_p / (kT_{sys})}{P_{RX} / (kT_{sys} B)} = B T_p \quad (23)$$

Thus, the final relation shows the SNR gain from matched filtering:

$$SNR_{after} = SNR_{before} BT_p \quad (24)$$

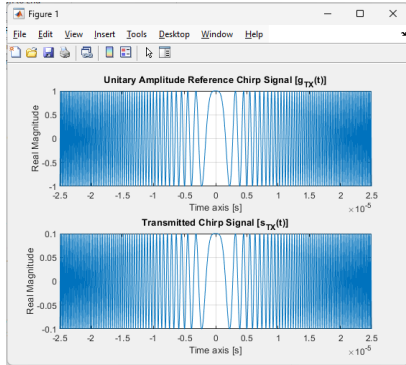
## 4 Matlab Simulation:

### 4.1 Given:

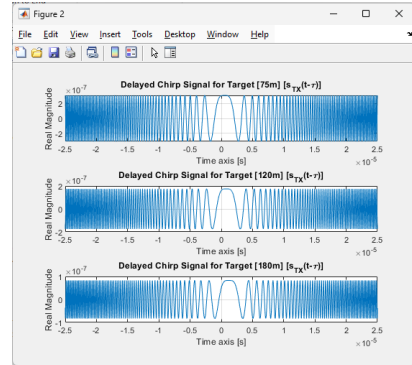
The steps to be performed in the simulation is given to us.

### 4.2 Results of the simulation:

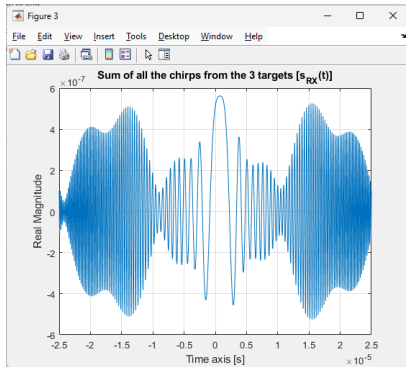
The results of the simulation is shown in Figure 3 and Figure 4.



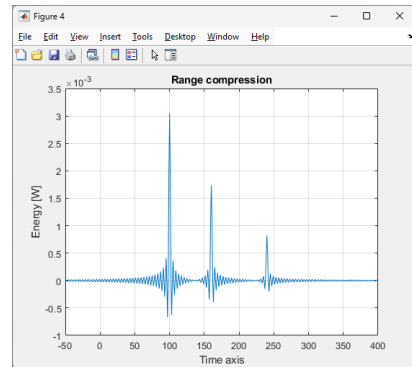
(a) Unit amplitude chirp and transmitted chirp signal



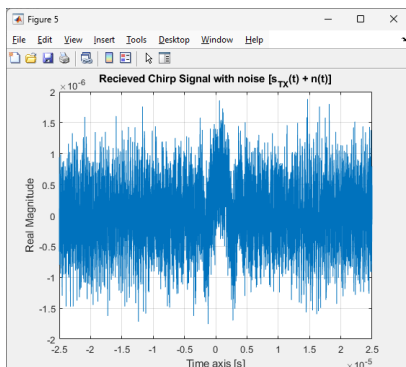
(b) Delayed, demodulated, attenuated signals



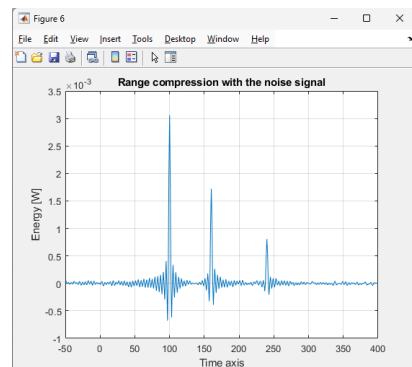
(c) Sum of the 3 reflected signals



(d) Range compression



(e) Sum of the 3 reflected signals with noise



(f) Range compression of the signals with noise

Figure 4: Results of the simulation

*Note: The Matlab Implementations are attached with the report.*